## **Supporting Information**

# Physical Reservoirs based on MoS<sub>2</sub>-HZO Integrated Ferroelectric Field-effect Transistors for Reservoir Computing Systems

Lingqi Li,<sup>a</sup> Heng Xiang,<sup>a</sup> Haofei Zheng,<sup>a</sup> Yu-Chieh Chien,<sup>a</sup> Ngoc Thanh Duong,<sup>a</sup> Jing Gao,<sup>a</sup> and Kah-Wee Ang<sup>\*a</sup> Department of Electrical and Computer Engineering, National University of Singapore, 4 Engineering Drive 3, 117583, Singapore E-mail: <u>eleakw@nus.edu.sg</u>



### S1. Demonstration of 1D-FeFET array

Fig. S1 (a) Microscopic image of the device array utilized in this study, with a scale

bar of 250  $\mu$ m. (b) Zoomed-in view highlighting the flake isolation.

#### S2. EDX results of MoS<sub>2</sub>-HZO FeFETs



Fig. S2 EDX map comprises W, Hf, Zr, O, Mo, and S elements.

#### **S3. FORC measurement schedule**



**Fig. S3** Electric field for the FORC measurement using MFM capacitor. The reversal electric field ( $E_r$ ) changes from 3.0 to -3.0 V with a step of -0.01 V, and the external field descends from 3.0 V to  $E_r$ , and then ascends to 3.0 V.



#### S4. Switching current for the NLS model fitting

**Fig. S4.** Switching current at different program voltage ( $V_P$ ) with varying programming duration ( $t_P$ ). (a) - (h) The  $V_P$  changes from 1.32 to 3.00 V, with a step of 0.24 V, and the  $t_P$  varies from 10 ns to 10 µs for each  $V_P$ . (i) Illustration of the switching current measurement setup.



**S5.** Deconvolution of varied results with different ozone pulse time

**Fig. S5** Deconvoluted XRD results in the 2 $\theta$  range of 27.5°-33.5° varied from  $t_{O3}$  of a) 0.1 s, b) 1.0 s, and c) 5.0 s.

#### S6. Long accumulation of response currents



**Fig. S6** Synaptic behavior by modulation of the conductance states with the pulse trains. Identical programming bias ( $V_P$ ) is introduced from 1.5 V to 3 V for all the input pulse scheme. The read voltage ( $V_R$ ) of 0.5 V with  $V_D$  = 0.5 V is fixed to all the input pulse trains. (a)  $t_{PW}$  = 100 µs,  $t_D$  = 100 µs (b)  $t_{PW}$  = 1 ms,  $t_D$  = 600 µs and (c)  $t_{PW}$  = 10 ms ,  $t_D$  = 600 µs.

One can observe that when we apply a short program pulse width  $(t_{PW})$ , the conductance difference  $(\Delta G_D)$  significantly increase with higher programming gate bias  $(V_P)$ . These observations indicating that more dipolar switching in the HZO

ferroelectric layer under larger voltage. On the other hand, when we further increase the  $t_{PW}$ , the  $\Delta G_D$  will not be directly proportional to the magnitude of  $V_P$ , which can be attributed to the serious trapping effect reduce the current. When we further increase the  $t_{PW}$ , the electron trapping become seriously which led the  $\Delta G_D$  smaller. This experimental result is consistent with previous reports <sup>1–4</sup> that the ferroelectric response is faster than the trapping response.





**Fig. S7.** (a) Energy band diagrams showing the band alignment of few-layer  $MoS_2$  to the high-k dielectric layer of  $Al_2O_3$  and  $HfO_2$ .<sup>5–7</sup> The accurate value of bandgaps, defect levels and band offsets were reported by several previous reports which extracted from the experiments. Transfer characteristics with f = 0.5 Hz at  $V_D$  = 50 mV and  $V_G$ from -1.0 V to 1.0 V for  $MoS_2$  FETs (b) with and (c) without 2 nm  $Al_2O_3$  interface layer on 10 nm HZO layer.

The transistor performance will be affected by the interfacial interaction between  $MoS_2$  and the high-k dielectric layer. In this work, we use a thin  $Al_2O_3$  layer (~2 nm) works as the interfacial layer between  $MoS_2$  and the ferroelectric HZO layer to minimize the detrimental charge trapping effect. As shown in the depicted energy

band diagrams (**Fig. S7a**), the acceptor-like traps (dark red) of HfO<sub>2</sub> are adjacent to the MoS<sub>2</sub> intrinsic Fermi level which energetically favorable to charge trapping effect. Compared with HfO<sub>2</sub>, the oxide defect states in the energy band of Al<sub>2</sub>O<sub>3</sub> shows the energy difference with the conduction band of MoS<sub>2</sub> which lead the detrimental charge trapping difficult and will lower the oxide trap density. Therefore, the depicted energy band diagrams confirm that the Al<sub>2</sub>O<sub>3</sub> align well with the MoS<sub>2</sub> so that suitable for the transistors.

Subsequently, the MoS<sub>2</sub> FETs with or without Al<sub>2</sub>O<sub>3</sub> interface layer are fabricated to verify the aforementioned deduce. Note that all the process except the Al<sub>2</sub>O<sub>3</sub> deposition was conducted together for these two samples. Then the same measurement scheme at  $V_D$  of 50 mV and the sweeping frequency (*f*) of 0.5 Hz was used for the devices. One can observe that the extracted clockwise hysteresis  $\Delta V_{TH}$ from the device with Al<sub>2</sub>O<sub>3</sub> interface layer is much smaller (~0.10V), while the  $\Delta V_{TH}$  is 0.26 V of the device without Al<sub>2</sub>O<sub>3</sub> over the same voltage range. Therefore, we have validated that address the interfacial interaction between MoS<sub>2</sub> and the high-k dielectric layer by Al<sub>2</sub>O<sub>3</sub> interface layer is suitability.



S8. MoS<sub>2</sub> FETs with pure Al<sub>2</sub>O<sub>3</sub> gate stack

**Fig. S8.** (a) Transfer characteristics and gate leakage current at  $V_D = 50$  mV and  $V_G$  from -2.0 V to 2.0 V for MoS<sub>2</sub> FETs using 12 nm pure Al<sub>2</sub>O<sub>3</sub> or 2 nm Al<sub>2</sub>O<sub>3</sub> stacks with 10nm HZO layer with *f* of 0.50 Hz. (b) The dynamic voltage response under a 3.0 V input pulse trains with  $t_{PW} = 100 \ \mu s$  and  $t_D = 100 \ \mu s$ .

It is evident that the device based on pure  $Al_2O_3$  shows a considerably smaller transient drain current compared to the HZO-based device, as shown in Fig. 2e. It confirms that the dynamic behavior observed is attributed to the ferroelectric HZO rather than the interfacial layer  $Al_2O_3$ . Furthermore, the smaller  $V_{TH}$  in our device can be attributed to enhanced electrical control.

## S9. CV characteristics of the HZO/Al<sub>2</sub>O<sub>3</sub> capacitor



**Fig. S9.** The frequency-dependent capacitance-voltage (CV) characteristics of  $HZO/Al_2O_3$  capacitor in similar MFM structure as in Fig. 1f. The thicknesses of HZO and  $Al_2O_3$  are 10 nm and 2 nm, respectively.

#### References

- K. Lee, H.-J. Lee, T. Y. Lee, H. H. Lim, M. S. Song, H. K. Yoo, D. I. Suh, J. G. Lee, Z. Zhu, A. Yoon, M. R. MacDonald, X. Lei, K. Park, J. Park, J. H. Lee and S. C. Chae, *ACS Appl. Mater. Interfaces*, 2019, **11**, 38929–38936.
- 2 M. Si, X. Lyu, P. R. Shrestha, X. Sun, H. Wang, K. P. Cheung and P. D. Ye, *Appl. Phys. Lett.*, 2019, **115**, 072107.
- 3 N. Gong, X. Sun, H. Jiang, K. S. Chang-Liao, Q. Xia and T. P. Ma, *Appl. Phys. Lett.*, 2018, **112**, 262903.
- 4 X. Lyu, M. Si, P. R. Shrestha, K. P. Cheung and P. D. Ye, in *2019 IEEE International Electron Devices Meeting (IEDM)*, 2019, p. 15.2.1-15.2.4.
- 5 Y. Y. Illarionov, T. Knobloch, M. Waltl, G. Rzepa, A. Pospischil, D. K. Polyushkin, M. M. Furchi, T. Mueller and T. Grasser, *2D Mater.*, 2017, *4*, 025108.
- 61. Shlyakhov, J. Chai, M. Yang, S. Wang, V. V. Afanas'ev, M. Houssa and A. Stesmans, *physica status solidi* (a), 2019, **216**, 1800616.
- 7 Y. Y. Illarionov, T. Knobloch, M. Jech, M. Lanza, D. Akinwande, M. I. Vexler, T. Mueller, M. C. Lemme, G. Fiori, F. Schwierz and T. Grasser, *Nat Commun*, 2020, **11**, 3385.